

National Semiconductor

LM2672 SIMPLE SWITCHER[®] Power Converter High Efficiency 1A Step-Down Voltage Regulator with Features

General Description

The LM2672 series of regulators are monolithic integrated circuits built with a LMDMOS process. These regulators provide all the active functions for a step-down (buck) switching regulator, capable of driving a 1A load current with excellent line and load regulation. These devices are available in fixed output voltages of 3.3V, 5.0V, 12V, and an adjustable output version.

Requiring a minimum number of external components, these regulators are simple to use and include patented internal frequency compensation (Patent Nos. 5,382,918 and 5,514,947), fixed frequency oscillator, external shutdown, soft-start, and frequency synchronization.

The LM2672 series operates at a switching frequency of 260 kHz, thus allowing smaller sized filter components than what would be needed with lower frequency switching regulators. Because of its very high efficiency (>90%), the copper traces on the printed circuit board are the only heat sinking needed.

A family of standard inductors for use with the LM2672 are available from several different manufacturers. This feature greatly simplifies the design of switch-mode power supplies using these advanced ICs. Also included in the datasheet are selector guides for diodes and capacitors designed to work in switch-mode power supplies.

Other features include a guaranteed $\pm 1.5\%$ tolerance on output voltage within specified input voltages and output load conditions, and $\pm 10\%$ on the oscillator frequency. External shutdown is included, featuring typically 50 µA stand-by current. The output switch includes current limiting, as well as thermal shutdown for full protection under fault conditions.

To simplify the LM2672 buck regulator design procedure, there exists computer design software, *LM267X Made Simple* version 6.0.

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Features

- Efficiency up to 96%
- Available in SO-8, 8-pin DIP and LLP packages
- Computer Design Software LM267X Made Simple version 6.0
- Simple and easy to design with
- Requires only 5 external components
- Uses readily available standard inductors
- 3.3V, 5.0V, 12V, and adjustable output versions
- Adjustable version output voltage range: 1.21V to 37V
- ±1.5% max output voltage tolerance over line and load conditions
- Guaranteed 1A output load current
- 0.25Ω DMOS Output Switch
- Wide input voltage range: 8V to 40V
- 260 kHz fixed frequency internal oscillator
- TTL shutdown capability, low power standby mode
- Soft-start and frequency synchronization
- Thermal shutdown and current limit protection

Typical Applications

- Simple High Efficiency (>90%) Step-Down (Buck) Regulator
- Efficient Pre-Regulator for Linear Regulators

Typical Application

(Fixed Output Voltage Versions)



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Connection Diagrams







LLP Package See NSC Package Drawing Number LDA16A

Output Voltage	Order Information	Package Marking	Supplied as:
6 Lead LLP			
12	LM2672LD-12	S0001B	1000 Units on Tape and Reel
12			4500 Units on Tape and Reel
3.3	LM2672LD-3.3	S0002B	1000 Units on Tape and Reel
3.3	LM2672LDX-3.3	S0002B	4500 Units on Tape and Reel
5.0	LM2672LD-5.0	S0003B	1000 Units on Tape and Reel
5.0	LM2672LDX-5.0	S0003B	4500 Units on Tape and Reel
ADJ	LM2672LD-ADJ	S0004B	1000 Units on Tape and Reel
ADJ	LM2672LDX-ADJ	S0004B	4500 Units on Tape and Reel
SO-8		2	
12	LM2672M-12	2672M-12	Shipped in Anti-Static Rails
12	LM2672MX-12	2672M-12	2500 Units on Tape and Reel
3.3	LM2672M-3.3	2672M-3.3	Shipped in Anti-Static Rails
3.3	LM2672MX-3.3	2672M-3.3	2500 Units on Tape and Reel
5.0	LM2672M-5.0	2672M-5.0	Shipped in Anti-Static Rails
5.0	LM2672MX-5.0	2672M-5.0	2500 Units on Tape and Ree
ADJ	LM2672M-ADJ	2672M-ADJ	Shipped in Anti-Static Rails
ADJ	LM2672MX-ADJ	2672M-ADJ	2500 Units on Tape and Reel
DIP			
12	LM2672N-12	LM2672N-12	Shipped in Anti-Static Rails
3.3	LM2672N-3.3	LM2672N-3.3	Shipped in Anti-Static Rails
5.0	LM2672N-5.0	LM2672N-5.0	Shipped in Anti-Static Rails
ADJ	LM2672N-ADJ	LM2672N-ADJ	Shipped in Anti-Static Rails

TABLE 1. Package Marking and Ordering Information

Absolute Maximum Ratings (Note 1)

If Military/Aerospace specified devices are required, please contact the National Semiconductor Sales Office/ Distributors for availability and specifications.

Supply Voltage	45V
ON/OFF Pin Voltage	$-0.1V \le V_{SH} \le 6V$
Switch Voltage to Ground	-1V
Boost Pin Voltage	V _{SW} + 8V
Feedback Pin Voltage	$-0.3V \le V_{FB} \le 14V$
ESD Susceptibility	
Human Body Model (Note 2)	2 kV
Power Dissipation	Internally Limited

Storage Temperature Range-65°C to +150°CLead TemperatureMM Package+215°CNafrared (15s)+220°CN Package (Soldering, 10s)+260°CLLP Package (see AN-1187)+150°CMaximum Junction Temperature+150°C

Operating Ratings

Supply Voltage	6.5V to 40V
Temperature Range	$-40^{\circ}C \le T_{J} \le +125^{\circ}C$

Electrical Characteristics

LM2672-3.3 Specifications with standard type face are for $T_J = 25^{\circ}$ C, and those in **bold type face** apply over **full Operating Temperature Range**.

Symbol	Parameter Conditions Typical (Note 4)		Min (Note 5)	Max (Note 5)	Units	
SYSTEM PARAMETERS Test Circuit Figure 2 (Note 3)						I
V _{OUT}	Output Voltage	$V_{IN} = 8V$ to 40V, $I_{LOAD} = 20$ mA to 1A	3.3	3.251/ 3.201	3.350/ 3.399	V
V _{OUT}	Output Voltage	V _{IN} = 6.5V to 40V, I _{LOAD} = 20 mA to 500 mA	3.3	3.251/ 3.201	3.350/ 3.399	V
η	Efficiency	$V_{IN} = 12V, I_{LOAD} = 1A$	86			%

LM2672-5.0

Symbol	Parameter	Conditions	Typical	Min	Max	Units
			(Note 4)	(Note 5)	(Note 5)	
SYSTEM PARAMETERS Test Circuit Figure 2 (Note 3)						
V _{OUT}	Output Voltage	$V_{IN} = 8V$ to 40V, $I_{LOAD} = 20$ mA to 1A	5.0	4.925/ 4.850	5.075/ 5.150	V
V _{OUT}	Output Voltage	$V_{IN} = 6.5V$ to 40V, $I_{LOAD} = 20$ mA to 500 mA	5.0	4.925/ 4.850	5.075/ 5.150	V
η	Efficiency	V _{IN} = 12V, I _{LOAD} = 1A	90			%

LM2672-12

Symbol	Parameter	Conditions Typ (No		Min (Note 5)	Max (Note 5)	Units	
SYSTEM PARAMETERS Test Circuit Figure 2 (Note 3)							
V _{OUT}	Output Voltage	$V_{IN} = 15V$ to 40V, $I_{LOAD} = 20$ mA to 1A	12	11.82/ 11.64	12.18/ 12.36	V	
η	Efficiency	$V_{IN} = 24V, I_{LOAD} = 1A$	94			%	

LM2672-ADJ

Symbol	Parameter	Conditions	Тур	Min	Max	Units
			(Note 4)	(Note 5)	(Note 5)	
SYSTEM	PARAMETERS Tes	st Circuit <i>Figure 3</i> (Note 3)		•		
V _{FB}	Feedback Voltage	$V_{IN} = 8V$ to 40V, $I_{LOAD} = 20$ mA to 1A	1.210	1.192/ 1.174	1.228/ 1.246	V
		V _{OUT} Programmed for 5V				
		(see Circuit of <i>Figure 3</i>)				
V _{FB}	Feedback Voltage	$V_{IN} = 6.5V$ to 40V, $I_{LOAD} = 20$ mA to 500 mA	1.210	1.192/ 1.174	1.228/ 1.246	V
		V _{OUT} Programmed for 5V				
		(see Circuit of <i>Figure 3</i>)				
η	Efficiency	$V_{IN} = 12V, I_{LOAD} = 1A$	90			%

All Output Voltage Versions

Specifications with standard type face are for $T_J = 25^{\circ}$ C, and those in **bold type face** apply over **full Operating Temperature Range**. Unless otherwise specified, $V_{IN} = 12$ V for the 3.3V, 5V, and Adjustable versions and $V_{IN} = 24$ V for the 12V version, and $I_{LOAD} = 100$ mA.

Symbol	Parameters	Conditions	Тур	Min	Max	Units
DEVICE	PARAMETERS	•	•		•	
Ι _Q	Quiescent Current	V _{FEEDBACK} = 8V For 3.3V, 5.0V, and ADJ Versions	2.5		3.6	mA
		V _{FEEDBACK} = 15V For 12V Versions	2.5			mA
I _{STBY}	Standby Quiescent Current	ON/OFF Pin = 0V	50		100/ 150	μA
I _{CL}	Current Limit		1.55	1.25/ 1.2	2.1/ 2.2	A
IL Output Leakage Current		$V_{IN} = 40V, ON/\overline{OFF}$ Pin = 0V $V_{SWITCH} = 0V$	1		25	μA
		$V_{SWITCH} = -1V$, ON/OFF Pin = 0V	6		15	mA
R _{DS(ON)}	Switch On-Resistance	I _{SWITCH} = 1A	0.25		0.30/ 0.50	Ω
f _o	Oscillator Frequency	Measured at Switch Pin	260	225	275	kHz
D	Maximum Duty Cycle		95			%
	Minimum Duty Cycle		0			%
I _{BIAS}	Feedback Bias Current	V _{FEEDBACK} = 1.3V ADJ Version Only	85			nA
V _{S/D}	ON/OFF Pin Voltage Thesholds		1.4	0.8	2.0	V
I _{S/D}	ON/OFF Pin Current	ON/OFF Pin = 0V	20	7	37	μA
F _{SYNC}	Synchronization Frequency	V _{SYNC} = 3.5V, 50% duty cycle	400			kHz
V _{SYNC}	Synchronization Threshold Voltage		1.4			v
V _{SS}	Soft-Start Voltage		0.63	0.53	0.73	V
I _{SS}	Soft-Start Current		4.5	1.5	6.9	μA
θ _{JA}	Thermal Resistance	N Package, Junction to Ambient (Note 6)	95			°C/W
		M Package, Junction to Ambient (Note 6)	105			

Note 1: Absolute Maximum Ratings indicate limits beyond which damage to the device may occur. Operating Ratings indicate conditions for which the device is intended to be functional, but device parameter specifications may not be guaranteed under these conditions. For guaranteed specifications and test conditions, see the Electrical Characteristics.

Note 2: The human body model is a 100 pF capacitor discharged through a 1.5 k Ω resistor into each pin.

Note 3: External components such as the catch diode, inductor, input and output capacitors, and voltage programming resistors can affect switching regulator performance. When the LM2672 is used as shown in *Figure 2* and *Figure 3* test circuits, system performance will be as specified by the system parameters section of the Electrical Characteristics.

Note 4: Typical numbers are at 25°C and represent the most likely norm.

Note 5: All limits guaranteed at room temperature (standard type face) and at temperature extremes (bold type face). All room temperature limits are 100% production tested. All limits at temperature extremes are guaranteed via correlation using standard Statistical Quality Control (SQC) methods. All limits are used to calculate Average Outgoing Quality Level (AOQL).

Note 6: Junction to ambient thermal resistance with approximately 1 square inch of printed circuit board copper surrounding the leads. Additional copper area will lower thermal resistance further. See Application Information section in the application note accompanying this datasheet and the thermal model in *LM267X Made Simple* version 6.0 software. The value θ_{J-A} for the LLP (LD) package is specifically dependent on PCB trace area, trace material, and the number of layers and thermal vias. For improved thermal resistance and power dissipation for the LLP package, refer to Application Note AN-1187.









INPUT VOLTAGE (V)

1293414

6





Typical Performance Characteristics (Circuit of Figure 2)



A: V_{SW} Pin Voltage, 10 V/div.







A: V_{SW} Pin Voltage, 10 V/div.

B: Inductor Current, 0.5 A/div

C: Output Ripple Voltage, 20 mV/div AC-Coupled

Horizontal Time Base: 1 µs/div



Load Transient Response for Discontinuous Mode V_{IN} = 20V, V_{OUT} = 5V, L = 47 µH, C_{OUT} = 68 µF, C_{OUT}ESR = 50 mΩ



A: Output Voltage, 100 mV/div, AC-Coupled B: Load Current: 100 mA to 300 mA Load Pulse

Horizontal Time Base: 200 µs/div



LM2672 Series Buck Regulator Design Procedure (Fixed Output)

PROCEDURE (Fixed Output Voltage Version) To simplify the buck regulator design procedure, National	EXAMPLE (Fixed Output Voltage Version)
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Semiconductor is making available computer design software to be	
used with the SIMPLE SWITCHER line of switching regulators.	
LM267X Made Simple version 6.0 is available on Windows [®] 3.1,	
NT, or 95 operating systems.	
Given:	Given:
V _{OUT} = Regulated Output Voltage (3.3V, 5V, or 12V)	$V_{OUT} = 5V$
V _{IN} (max) = Maximum DC Input Voltage	$V_{\rm IN}(\rm max) = 12V$
I _{LOAD} (max) = Maximum Load Current	$I_{LOAD}(max) = 1A$
1. Inductor Selection (L1)	1. Inductor Selection (L1)
A. Select the correct inductor value selection guide from <i>Figure 4</i>	A. Use the inductor selection guide for the 5V version shown in
and <i>Figure 5</i> or <i>Figure 6</i> (output voltages of 3.3V, 5V, or 12V	Figure 5.
respectively). For all other voltages, see the design procedure for	
the adjustable version.	
B. From the inductor value selection guide, identify the inductance	B. From the inductor value selection guide shown in <i>Figure 5</i> , the
region intersected by the Maximum Input Voltage line and the	inductance region intersected by the 12V horizontal line and the 1A
Maximum Load Current line. Each region is identified by an	vertical line is 33 μ H, and the inductor code is L23.
inductance value and an inductor code (LXX).	
$\ensuremath{\textbf{C}}\xspace$. Select an appropriate inductor from the four manufacturer's part	C. The inductance value required is 33 μ H. From the table in <i>Figure</i>
numbers listed in Figure 8. Each manufacturer makes a different	8, go to the L23 line and choose an inductor part number from any
style of inductor to allow flexibility in meeting various design	of the four manufacturers shown. (In most instances, both through
requirements. Listed below are some of the differentiating	hole and surface mount inductors are available.)
characteristics of each manufacturer's inductors:	
Schott: ferrite EP core inductors; these have very low leakage	
magnetic fields to reduce electro-magnetic interference (EMI) and	
are the lowest power loss inductors	
<i>Renco:</i> ferrite stick core inductors; benefits are typically lowest cost inductors and can withstand E•T and transient peak currents above	
rated value. Be aware that these inductors have an external	
magnetic field which may generate more EMI than other types of	
inductors.	
<i>Pulse:</i> powered iron toroid core inductors; these can also be low	
cost and can withstand larger than normal E•T and transient peak	
currents. Toroid inductors have low EMI.	
Coilcraft: ferrite drum core inductors; these are the smallest	
physical size inductors, available only as SMT components. Be	
aware that these inductors also generate EMI-but less than stick	
inductors.	
Complete specifications for these inductors are available from the	
respective manufacturers. A table listing the manufacturers' phone	
numbers is located in Figure 9.	
2. Output Capacitor Selection (C _{OUT})	2. Output Capacitor Selection (C _{OUT})
A. Select an output capacitor from the output capacitor table in	A. Use the 5.0V section in the output capacitor table in <i>Figure 10</i> .
Figure 10. Using the output voltage and the inductance value found	Choose a capacitor value and voltage rating from the line that
in the inductor selection guide, step 1, locate the appropriate	contains the inductance value of 33 μ H. The capacitance and
capacitor value and voltage rating.	voltage rating values corresponding to the 33 μ H

PROCEDURE (Fixed Output Voltage Version)	EXAMPLE (Fixed Output Voltage Version)
The capacitor list contains through-hole electrolytic capacitors from	Surface Mount:
four different capacitor manufacturers and surface mount tantalum	68 μF/10V Sprague 594D Series.
capacitors from two different capacitor manufacturers. It is	100 μF/10V AVX TPS Series.
recommended that both the manufacturers and the manufacturer's	Through Hole:
series that are listed in the table be used. A table listing the	68 μF/10V Sanyo OS-CON SA Series.
manufacturers' phone numbers is located in <i>Figure 11</i> .	220 μF/35V Sanyo MV-GX Series.
	220 µF/35V Nichicon PL Series.
	220 µF/35V Panasonic HFQ Series.
3. Catch Diode Selection (D1)	3. Catch Diode Selection (D1)
A. In normal operation, the average current of the catch diode is	A. Refer to the table shown in <i>Figure 12</i> . In this example, a 1A,
the load current times the catch diode duty cycle, 1-D (D is the	20V Schottky diode will provide the best performance. If the circu
switch duty cycle, which is approximately the output voltage divided	must withstand a continuous shorted output, a higher current
by the input voltage). The largest value of the catch diode average	Schottky diode is recommended.
current occurs at the maximum load current and maximum input	
voltage (minimum D). For normal operation, the catch diode current	
rating must be at least 1.3 times greater than its maximum average	
current. However, if the power supply design must withstand a	
continuous output short, the diode should have a current rating	
equal to the maximum current limit of the LM2672. The most	
stressful condition for this diode is a shorted output condition.	
B. The reverse voltage rating of the diode should be at least 1.25	
times the maximum input voltage.	
C. Because of their fast switching speed and low forward voltage	
drop, Schottky diodes provide the best performance and efficiency.	
This Schottky diode must be located close to the LM2672 using	
short leads and short printed circuit traces.	
4. Input Capacitor (C _{IN})	4. Input Capacitor (C _{IN})
A low ESR aluminum or tantalum bypass capacitor is needed	The important parameters for the input capacitor are the input
between the input pin and ground to prevent large voltage	voltage rating and the RMS current rating. With a maximum input
transients from appearing at the input. This capacitor should be	voltage of 12V, an aluminum electrolytic capacitor with a voltage
located close to the IC using short leads. In addition, the RMS	rating greater than 15V (1.25 \times V _{IN}) would be needed. The next
current rating of the input capacitor should be selected to be at least	higher capacitor voltage rating is 16V.
1/2 the DC load current. The capacitor manufacturer data sheet must	The RMS current rating requirement for the input capacitor in a
be checked to assure that this current rating is not exceeded. The	buck regulator is approximately 1/2 the DC load current. In this
curves shown in <i>Figure 14</i> show typical RMS current ratings for	example, with a 1A load, a capacitor with a RMS current rating of
several different aluminum electrolytic capacitor values. A parallel	at least 500 mA is needed. The curves shown in Figure 14 can be
connection of two or more capacitors may be required to increase	used to select an appropriate input capacitor. From the curves,
the total minimum RMS current rating to suit the application	locate the 16V line and note which capacitor values have RMS
requirements.	current ratings greater than 500 mA.
For an aluminum electrolytic capacitor, the voltage rating should be	For a through hole design, a 330 µF/16V electrolytic capacitor
at least 1.25 times the maximum input voltage. Caution must be	(Panasonic HFQ series, Nichicon PL, Sanyo MV-GX series or
exercised if solid tantalum capacitors are used. The tantalum	equivalent) would be adequate. Other types or other
capacitor voltage rating should be twice the maximum input	manufacturers' capacitors can be used provided the RMS ripple
voltage. The tables in Figure 15 show the recommended	current ratings are adequate. Additionally, for a complete surface
application voltage for AVX TPS and Sprague 594D tantalum	mount design, electrolytic capacitors such as the Sanyo CV-C or
capacitors. It is also recommended that they be surge current	CV-BS and the Nichicon WF or UR and the NIC Components NAC
	series could be considered.
tested by the manufacturer. The TPS series available from AVX,	For surface mount designs, solid tantalum capacitors can be used
and the 593D and 594D series from Sprague are all surge current	i or surface mount designs, solid tantalum capacitors can be used
and the 593D and 594D series from Sprague are all surge current tested. Another approach to minimize the surge current stresses	
and the 593D and 594D series from Sprague are all surge current tested. Another approach to minimize the surge current stresses on the input capacitor is to add a small inductor in series with the	but caution must be exercised with regard to the capacitor surge
and the 593D and 594D series from Sprague are all surge current tested. Another approach to minimize the surge current stresses on the input capacitor is to add a small inductor in series with the input supply line.	
and the 593D and 594D series from Sprague are all surge current tested. Another approach to minimize the surge current stresses on the input capacitor is to add a small inductor in series with the	but caution must be exercised with regard to the capacitor surge current rating and voltage rating. In this example, checking <i>Figure</i>

PROCEDURE (Fixed Output Voltage Version)	EXAMPLE (Fixed Output Voltage Version)
5. Boost Capacitor (C _B)	5. Boost Capacitor (C _B)
This capacitor develops the necessary voltage to turn the switch	For this application, and all applications, use a 0.01 μ F, 50V
gate on fully. All applications should use a 0.01 µF, 50V ceramic	ceramic capacitor.
capacitor.	
6. Soft-Start Capacitor (C _{ss} - optional)	6. Soft-Start Capacitor (C _{SS} - optional)
This capacitor controls the rate at which the device starts up. The	For this application, selecting a start-up time of 10 ms and usir
formula for the soft-start capacitor C_{SS} is:	the formula for C_{SS} results in a value of:
$C_{SS} \approx (I_{SS} \cdot t_{SS}) / [V_{SSTH} + 2.6V \cdot (\frac{V_{OUT} + V_{SCHOTTKY}}{V_{IN}})]$	C _{SS} ≈ (4.5 μ A • 10 ms) / [0.63V + 2.6V • ($\frac{5V + 0.4V}{12V}$)
IN	= 25 nF ≈ 0.022 μF.
where:	
I _{SS} = Soft-Start Current :4.5 μA typical.	
t_{SS} = Soft-Start Time :Selected.	
V_{SSTH} = Soft-Start Threshold Voltage :0.63V typical.	
V_{SSTH} = Solitotari Theshold Voltage .0.00V typical. V_{OUT} = Output Voltage :Selected.	
V _{SCHOTTKY} = Schottky Diode Voltage Drop :0.4V typical.	
V _{IN} = Input Voltage :Selected.	
If this feature is not desired, leave this pin open. With certain	
softstart capacitor values and operating conditions, the LM2672	
can exhibit an overshoot on the output voltage during turn on.	
Especially when starting up into no load or low load, the softstart	
function may not be effective in preventing a larger voltage	
overshoot on the output. With larger loads or lower input voltages	
during startup this effect is minimized. In particular, avoid using	
softstart capacitors between 0.033µF and 1µF.	
7. Frequency Synchronization (optional)	7. Frequency Synchronization (optional)
The LM2672 (oscillator) can be synchronized to run with an	For all applications, use a 1 k Ω resistor and a 100 pF capacitor
external oscillator, using the sync pin (pin 3). By doing so, the	the RC filter.
LM2672 can be operated at higher frequencies than the standard	
frequency of 260 kHz. This allows for a reduction in the size of the	
inductor and output capacitor.	
As shown in the drawing below, a signal applied to a RC filter at the	
sync pin causes the device to synchronize to the frequency of that	
signal. For a signal with a peak-to-peak amplitude of $\ensuremath{3V}$ or greater,	
a 1 k Ω resistor and a 100 pF capacitor are suitable values.	
$V_{S \ge 3.0V} \qquad I \qquad O \qquad I \qquad SYNC \qquad LM2672$	
°↓ ¨ L	

Inductor Value Selection Guides

(For Continuous Mode Operation)



FIGURE 4. LM2672-3.3



FIGURE 5. LM2672-5.0



FIGURE 6. LM2672-12



FIGURE 7. LM2672-ADJ

Ind.	Inductanc	0	Sch	nott	Rend	0	Pulse Er	ngineering	Coilcraft
Ref.	е	Current	Through	Surface	Through	Surface	Through	Surface	Surface
Desg.	(µH)	(A)	Hole	Mount	Hole	Mount	Hole	Mount	Mount
L4	68	0.32	67143940	67144310	RL-1284-68-43	RL1500-68	PE-53804	PE-53804-S	DO1608-683
L5	47	0.37	67148310	67148420	RL-1284-47-43	RL1500-47	PE-53805	PE-53805-S	DO1608-473
L6	33	0.44	67148320	67148430	RL-1284-33-43	RL1500-33	PE-53806	PE-53806-S	DO1608-333
L7	22	0.52	67148330	67148440	RL-1284-22-43	RL1500-22	PE-53807	PE-53807-S	DO1608-223
L9	220	0.32	67143960	67144330	RL-5470-3	RL1500-220	PE-53809	PE-53809-S	DO3308-224
L10	150	0.39	67143970	67144340	RL-5470-4	RL1500-150	PE-53810	PE-53810-S	DO3308-154
L11	100	0.48	67143980	67144350	RL-5470-5	RL1500-100	PE-53811	PE-53811-S	DO3308-104
L12	68	0.58	67143990	67144360	RL-5470-6	RL1500-68	PE-53812	PE-53812-S	DO3308-683
L13	47	0.70	67144000	67144380	RL-5470-7	RL1500-47	PE-53813	PE-53813-S	DO3308-473
L14	33	0.83	67148340	67148450	RL-1284-33-43	RL1500-33	PE-53814	PE-53814-S	DO3308-333
L15	22	0.99	67148350	67148460	RL-1284-22-43	RL1500-22	PE-53815	PE-53815-S	DO3308-223
L18	220	0.55	67144040	67144420	RL-5471-2	RL1500-220	PE-53818	PE-53818-S	DO3316-224
L19	150	0.66	67144050	67144430	RL-5471-3	RL1500-150	PE-53819	PE-53819-S	DO3316-154
L20	100	0.82	67144060	67144440	RL-5471-4	RL1500-100	PE-53820	PE-53820-S	DO3316-104
L21	68	0.99	67144070	67144450	RL-5471-5	RL1500-68	PE-53821	PE-53821-S	DO3316-683
L22	47	1.17	67144080	67144460	RL-5471-6		PE-53822	PE-53822-S	DO3316-473
L23	33	1.40	67144090	67144470	RL-5471-7	_	PE-53823	PE-53823-S	DO3316-333

Ind. Inductanc		•	Schott		Renco		Pulse Engineering		Coilcraft
Ref.	е	Current	Through	Surface	Through	Surface	Through	Surface	Surface
Desg.	(µH)	(A)	Hole	Mount	Hole	Mount	Hole	Mount	Mount
L24	22	1.70	67148370	67148480	RL-1283-22-43	_	PE-53824	PE-53824-S	DO3316-223
L27	220	1.00	67144110	67144490	RL-5471-2	_	PE-53827	PE-53827-S	DO5022P-224
L28	150	1.20	67144120	67144500	RL-5471-3	_	PE-53828	PE-53828-S	DO5022P-154
L29	100	1.47	67144130	67144510	RL-5471-4	_	PE-53829	PE-53829-S	DO5022P-104
L30	68	1.78	67144140	67144520	RL-5471-5	_	PE-53830	PE-53830-S	DO5022P-683

FIGURE 8. Inductor Manufacturers' Part Numbers

Coilcraft Inc.	Phone	(800) 322-2645
		· · ·
	FAX	(708) 639-1469
Coilcraft Inc., Europe	Phone	+44 1236 730 595
	FAX	+44 1236 730 627
Pulse Engineering Inc.	Phone	(619) 674-8100
	FAX	(619) 674-8262
Pulse Engineering Inc.,	Phone	+353 93 24 107
Europe	FAX	+353 93 24 459
Renco Electronics Inc.	Phone	(800) 645-5828
	FAX	(516) 586-5562
Schott Corp.	Phone	(612) 475-1173
	FAX	(612) 475-1786

FIGURE 9. Inductor Manufacturers' Phone Numbers

		Output Capacitor					
Output	Inductance (µH)	Surface Mount		Through Hole			
Voltage		Sprague	AVX TPS	Sanyo OS-CON	Sanyo MV-GX	Nichicon	Panasonic
(V)		594D Series	Series	SA Series	Series	PL Series	HFQ Series
		(µF/V)	(µF/V)	(µF/V)	(µF/V)	(µF/V)	(µF/V)
	22	120/6.3	100/10	100/10	330/35	330/35	330/35
	33	120/6.3	100/10	68/10	220/35	220/35	220/35
3.3	47	68/10	100/10	68/10	150/35	150/35	150/35
3.3	68	120/6.3	100/10	100/10	120/35	120/35	120/35
	100	120/6.3	100/10	100/10	120/35	120/35	120/35
	150	120/6.3	100/10	100/10	120/35	120/35	120/35
	22	100/16	100/10	100/10	330/35	330/35	330/35
	33	68/10	10010	68/10	220/35	220/35	220/35
5.0	47	68/10	100/10	68/10	150/35	150/35	150/35
5.0	68	100/16	100/10	100/10	120/35	120/35	120/35
	100	100/16	100/10	100/10	120/35	120/35	120/35
	150	100/16	100/10	100/10	120/35	120/35	120/35
	22	120/20	(2×) 68/20	68/20	330/35	330/35	330/35
	33	68/25	68/20	68/20	220/35	220/35	220/35
	47	47/20	68/20	47/20	150/35	150/35	150/35
12	68	47/20	68/20	47/20	120/35	120/35	120/35
	100	47/20	68/20	47/20	120/35	120/35	120/35
	150	47/20	68/20	47/20	120/35	120/35	120/35
	220	47/20	68/20	47/20	120/35	120/35	120/35

FIGURE 10. Output Capacitor Table

Nichicon Corp.	Phone	(847) 843-7500
	FAX	(847) 843-2798
Panasonic	Phone	(714) 373-7857
	FAX	(714) 373-7102
AVX Corp.	Phone	(803) 448-9411
	FAX	(803) 448-1943
Sprague/Vishay	Phone	(207) 324-4140
	FAX	(207) 324-7223
Sanyo Corp.	Phone	(619) 661-6322
	FAX	(619) 661-1055

FIGURE 11. Capacitor Manufacturers' Phone Numbers

	1A Di	odes	3A Di	liodes	
V _R	Surface	Through	Surface	Through	
	Mount	Hole	Mount	Hole	
20V	SK12	1N5817	SK32	1N5820	
	B120	SR102		SR302	
30V	SK13	1N5818	SK33	1N5821	
	B130	11DQ03	30WQ03F	31DQ03	
	MBRS130	SR103			
40V	SK14	1N5819	SK34	1N5822	
	B140	11DQ04	30BQ040	MBR340	
	MBRS140	SR104	30WQ04F	31DQ04	
	10BQ040		MBRS340	SR304	
	10MQ040		MBRD340		
	15MQ040				
50V	SK15	MBR150	SK35	MBR350	
	B150	11DQ05	30WQ05F	31DQ05	
	10BQ050	SR105		SR305	

FIGURE 12. Schottky Diode Selection Table

International Rectifier Corp.	Phone	(310) 322-3331
	FAX	(310) 322-3332
Motorola, Inc.	Phone	(800) 521-6274
	FAX	(602) 244-6609
General Instruments Corp.	Phone	(516) 847-3000
	FAX	(516) 847-3236
Diodes, Inc.	Phone	(805) 446-4800
	FAX	(805) 446-4850

FIGURE 13. Diode Manufacturers' Phone Numbers





Recommended Application Voltage +85°C Ratii	Voltage Rating ng
3.3	6.3
5	10
10	20
12	25
15	35

Sprague 594D

Recommended Application Voltage	Voltage Rating			
+85°C Rating				
2.5	4			
3.3	6.3			
5	10			
8	16			
12	20			
18	25			
24	35			
29	50			

FIGURE 15. Recommended Application Voltage for AVX TPS and Sprague 594D Tantalum Chip Capacitors Derated for 85°C.

LM2672 Series Buck Regulator Design Procedure (Adjustable Output)

PROCEDURE (Adjustable Output Voltage Version)	EXAMPLE (Adjustable Output Voltage Version)
To simplify the buck regulator design procedure, National	
Semiconductor is making available computer design software to be	
used with the SIMPLE SWITCHER line of switching regulators.	
LM267X Made Simple version 6.0 is available on Windows 3.1,	
NT, or 95 operating systems.	
Given:	Given:
V _{OUT} = Regulated Output Voltage	$V_{OUT} = 20V$
V _{IN} (max) = Maximum Input Voltage	$V_{\rm IN}(\rm max) = 28V$
	$I_{LOAD}(max) = 1A$
I _{LOAD} (max) = Maximum Load Current	
F = Switching Frequency (<i>Fixed at a nominal 260 kHz</i>).	F = Switching Frequency (<i>Fixed at a nominal 260 kHz</i>).
1. Programming Output Voltage (Selecting R_1 and R_2 , as shown	
in <i>Figure 3</i>)	in <i>Figure 3</i>)
Use the following formula to select the appropriate resistor values.	Select R_1 to be 1 k Ω , 1%. Solve for R_2 .
$V_{OUT} = V_{REF} \left(1 + \frac{R_2}{R_1} \right)$	$R_{2} = R_{1} \left(\frac{V_{OUT}}{V_{RFF}} - 1 \right) = 1 k\Omega \left(\frac{20V}{1.23V} - 1 \right)$
where $V_{REF} = 1.21V$	
Select a value for $R^{}_1$ between 240 Ω and 1.5 k $\Omega.$ The lower resistor	
values minimize noise pickup in the sensitive feedback pin. (For the lowest temperature coefficient and the best stability with time, use	$R_2 = 15.4 \text{ k}\Omega.$
1% metal film resistors.)	
$R_{2} = R_{1} \left(\frac{V_{OUT}}{V_{REF}} - 1 \right)$	
2. Inductor Selection (L1)	2. Inductor Selection (L1)
A. Calculate the inductor Volt • microsecond constant E • T	A. Calculate the inductor Volt • microsecond constant (E • T),
(V • μs), from the following formula:	
$E \cdot T = (V_{IN(MAX)} - V_{OUT} - V_{SAT}) \cdot \frac{V_{OUT} + V_D}{V_{IN(MAX)} - V_{SAT} + V_D} \cdot \frac{1000}{260} (V \cdot \mu s)$	$E \cdot T = (28 - 20 - 0.25) \cdot \frac{20 + 0.5}{28 - 0.25 + 0.5} \cdot \frac{1000}{260} (V \cdot \mu s)$
	$E \cdot T = (7.75) \cdot \frac{20.5}{28.25} \cdot 3.85 (V \cdot \mu s) = 21.6 (V \cdot \mu s)$
where V _{SAT} =internal switch saturation voltage=0.25V and	
$V_{\rm D}$ = diode forward voltage drop = 0.5V	
B. Use the E • T value from the previous formula and match it with	B. E • T = 21.6 (V • μs)
the E • T number on the vertical axis of the Inductor Value Selection	
Guide shown in Figure 7.	
C. On the horizontal axis, select the maximum load current.	C. $I_{LOAD}(max) = 1A$
D. Identify the inductance region intersected by the E • T value and	
the Maximum Load Current value. Each region is identified by an	inductance region intersected by the 21.6 (V \bullet µs) horizontal line
inductance value and an inductor code (LXX).	and the 1A vertical line is 68 μ H, and the inductor code is L30.
E. Select an appropriate inductor from the four manufacturer's part	-
numbers listed in <i>Figure 8</i> . For information on the different types of	_
inductors, see the inductor selection in the fixed output voltage	
design procedure.	
3. Output Capacitor Selection (C _{OUT})	3. Output Capacitor Selection (C _{OUT})
	A. Use the appropriate row of the capacitor code selection guide
A. Select an output capacitor from the capacitor code selection	
A. Select an output capacitor from the capacitor code selection quide in <i>Figure 16</i> . Using the inductance value found in the inductor	I in <i>Figure 16</i> . For this example, use the 15–20V row. The capacito
guide in Figure 16. Using the inductance value found in the inductor	in <i>Figure 16</i> . For this example, use the 15–20V row. The capacito code corresponding to an inductance of 68 µH is C20.
	in <i>Figure 16</i> . For this example, use the 15–20V row. The capacito code corresponding to an inductance of 68 μH is C20.

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PROCEDURE (Adjustable Output Voltage Version)	EXAMPLE (Adjustable Output Voltage Version)
B. Select an appropriate capacitor value and voltage rating, using	B. From the output capacitor selection table in Figure 17, choose
the capacitor code, from the output capacitor selection table in	a capacitor value (and voltage rating) that intersects the capacitor
Figure 17. There are two solid tantalum (surface mount) capacitor	code(s) selected in section A, C20.
manufacturers and four electrolytic (through hole) capacitor	The capacitance and voltage rating values corresponding to the
manufacturers to choose from. It is recommended that both the	capacitor code C20 are the:
manufacturers and the manufacturer's series that are listed in the	Surface Mount:
table be used. A table listing the manufacturers' phone numbers is	33 μF/25V Sprague 594D Series.
located in Figure 11.	33 µF/25V AVX TPS Series.
	Through Hole:
	33 μF/25V Sanyo OS-CON SC Series.
	120 µF/35V Sanyo MV-GX Series.
	120 µF/35V Nichicon PL Series.
	120 µF/35V Panasonic HFQ Series.
	Other manufacturers or other types of capacitors may also be used
	provided the capacitor specifications (especially the 100 kHz ESF
	closely match the characteristics of the capacitors listed in the
	output capacitor table. Refer to the capacitor manufacturers' data
	sheet for this information.
4. Catch Diode Selection (D1)	4. Catch Diode Selection (D1)
A. In normal operation, the average current of the catch diode is	A. Refer to the table shown in Figure 12. Schottky diodes provid
the load current times the catch diode duty cycle, 1-D (D is the	the best performance, and in this example a 1A, 40V Schottky diod
switch duty cycle, which is approximately V _{OUT} /V _{IN}). The largest	would be a good choice. If the circuit must withstand a continuou
value of the catch diode average current occurs at the maximum	shorted output, a higher current (at least 2.2A) Schottky diode is
input voltage (minimum D). For normal operation, the catch diode	recommended.
current rating must be at least 1.3 times greater than its maximum	
average current. However, if the power supply design must	
withstand a continuous output short, the diode should have a	
current rating greater than the maximum current limit of the	

output condition. **B.** The reverse voltage rating of the diode should be at least 1.25 times the maximum input voltage.

LM2672. The most stressful condition for this diode is a shorted

C. Because of their fast switching speed and low forward voltage drop, Schottky diodes provide the best performance and efficiency. The Schottky diode must be located close to the LM2672 using short leads and short printed circuit traces.

PROCEDURE (Adjustable Output Voltage Version)	EXAMPLE (Adjustable Output Voltage Version)				
5. Input Capacitor (C _{IN})	5. Input Capacitor (C _{IN})				
A low ESR aluminum or tantalum bypass capacitor is needed between the input pin and ground to prevent large voltage transients from appearing at the input. This capacitor should be located close to the IC using short leads. In addition, the RMS current rating of the input capacitor should be selected to be at least ½ the DC load current. The capacitor manufacturer data sheet must be checked to assure that this current rating is not exceeded. The curves shown in <i>Figure 14</i> show typical RMS current ratings for several different aluminum electrolytic capacitor values. A parallel connection of two or more capacitors may be required to increase the total minimum RMS current rating to suit the application requirements. For an aluminum electrolytic capacitor, the voltage rating should be at least 1.25 times the maximum input voltage. Caution must be	The important parameters for the input capacitor are the input voltage rating and the RMS current rating. With a maximum input voltage of 28V, an aluminum electrolytic capacitor with a voltage rating of at least 35V ($1.25 \times V_{IN}$) would be needed. The RMS current rating requirement for the input capacitor in a buck regulator is approximately ½ the DC load current. In this example, with a 1A load, a capacitor with a RMS current rating of at least 500 mA is needed. The curves shown in <i>Figure 14</i> can be used to select an appropriate input capacitor. From the curves, locate the 35V line and note which capacitor values have RMS current ratings greater than 500 mA. For a through hole design, a 330 µF/35V electrolytic capacitor (Panasonic HFQ series, Nichicon PL, Sanyo MV-GX series or equivalent) would be adequate. Other types or other				
exercised if solid tantalum capacitors are used. The tantalum capacitor voltage rating should be twice the maximum input voltage. The tables in <i>Figure 15</i> show the recommended application voltage for AVX TPS and Sprague 594D tantalum capacitors. It is also recommended that they be surge current tested by the manufacturer. The TPS series available from AVX, and the 593D and 594D series from Sprague are all surge current tested. Another approach to minimize the surge current stresses on the input capacitor is to add a small inductor in series with the input supply line. Use caution when using ceramic capacitors for input bypassing, because it may cause severe ringing at the V _{IN} pin.	manufacturers' capacitors can be used provided the RMS ripple current ratings are adequate. Additionally, for a complete surface mount design, electrolytic capacitors such as the Sanyo CV-C or CV-BS and the Nichicon WF or UR and the NIC Components NACZ series could be considered. For surface mount designs, solid tantalum capacitors can be used, but caution must be exercised with regard to the capacitor surge current rating and voltage rating. In this example, checking <i>Figure</i> <i>15</i> , and the Sprague 594D series datasheet, a Sprague 594D 15 μ F, 50V capacitor is adequate.				
6. Boost Capacitor (C_B) This capacitor develops the necessary voltage to turn the switch gate on fully. All applications should use a 0.01 μ F, 50V ceramic capacitor.	6. Boost Capacitor (C_B) For this application, and all applications, use a 0.01 μ F, 50V ceramic capacitor.				

If the soft-start and frequency synchronization features are desired, look at steps 6 and 7 in the fixed output design procedure.

Case	Output	Inductance (µH)							
Style (Note 7)	Voltage (V)	22	33	47	68	100	150	220	
SM and TH	1.21–2.50	—	_	_	_	C1	C2	C3	
SM and TH	2.50–3.75	_	_	_	C1	C2	C3	C3	
SM and TH	3.75–5.0	_	_	C4	C5	C6	C6	C6	
SM and TH	5.0–6.25	_	C4	C7	C6	C6	C6	C6	
SM and TH	6.25–7.5	C8	C4	C7	C6	C6	C6	C6	
SM and TH	7.5–10.0	C9	C10	C11	C12	C13	C13	C13	
SM and TH	10.0–12.5	C14	C11	C12	C12	C13	C13	C13	
SM and TH	12.5–15.0	C15	C16	C17	C17	C17	C17	C17	
SM and TH	15.0–20.0	C18	C19	C20	C20	C20	C20	C20	
SM and TH	20.0–30.0	C21	C22	C22	C22	C22	C22	C22	
TH	30.0–37.0	C23	C24	C24	C25	C25	C25	C25	

Note 7: SM - Surface Mount, TH - Through Hole

FIGURE 16. Capacitor Code Selection Guide

			Output Capacite	or		
Cap.	Surface	Mount	Through Hole			
Ref. Desq.	Sprague 594D Series	AVX TPS Series	Sanyo OS-CON SA Series	Sanyo MV-GX Series	Nichicon PL Series	Panasonic HFQ Series
#	(µF/V)	(µF/V)	(µF/V)	(µF/V)	(µF/V)	(µF/V)
C1	120/6.3	100/10	100/10	220/35	220/35	220/35
C2	120/6.3	100/10	100/10	150/35	150/35	150/35
C3	120/6.3	100/10	100/35	120/35	120/35	120/35
C4	68/10	100/10	68/10	220/35	220/35	220/35
C5	100/16	100/10	100/10	150/35	150/35	150/35
C6	100/16	100/10	100/10	120/35	120/35	120/35
C7	68/10	100/10	68/10	150/35	150/35	150/35
C8	100/16	100/10	100/10	330/35	330/35	330/35
C9	100/16	100/16	100/16	330/35	330/35	330/35
C10	100/16	100/16	68/16	220/35	220/35	220/35
C11	100/16	100/16	68/16	150/35	150/35	150/35
C12	100/16	100/16	68/16	120/35	120/35	120/35
C13	100/16	100/16	100/16	120/35	120/35	120/35
C14	100/16	100/16	100/16	220/35	220/35	220/35
C15	47/20	68/20	47/20	220/35	220/35	220/35
C16	47/20	68/20	47/20	150/35	150/35	150/35
C17	47/20	68/20	47/20	120/35	120/35	120/35
C18	68/25	(2×) 33/25	47/25 (Note 8)	220/35	220/35	220/35
C19	33/25	33/25	33/25 (Note 8)	150/35	150/35	150/35
C20	33/25	33/25	33/25 (Note 8)	120/35	120/35	120/35
C21	33/35	(2×) 22/25	(Note 9)	150/35	150/35	150/35
C22	33/35	22/35	(Note 9)	120/35	120/35	120/35
C23	(Note 9)	(Note 9)	(Note 9)	220/50	100/50	120/50
C24	(Note 9)	(Note 9)	(Note 9)	150/50	100/50	120/50
C25	(Note 9)	(Note 9)	(Note 9)	150/50	82/50	82/50

Note 8: The SC series of Os-Con capacitors (others are SA series)

Note 9: The voltage ratings of the surface mount tantalum chip and Os-Con capacitors are too low to work at these voltages.

FIGURE 17. Output Capacitor Selection Table

Application Information

TYPICAL SURFACE MOUNT PC BOARD LAYOUT, FIXD OUTPUT (4X SIZE)



$$\begin{split} &C_{IN} - 15 \, \mu F, \, 50V, \, Solid \, Tantalum \, Sprague, \, "594D \, series" \\ &C_{OUT} - 68 \, \mu F, \, 16V, \, Solid \, Tantalum \, Sprague, \, "594D \, series" \\ &D1 - 1A, \, 40V \, Schottky \, Rectifier, \, Surface \, Mount \end{split}$$

L1 - 33 µH, L23, Coilcraft DO3316

 $C_B - 0.01 \ \mu\text{F}$, 50V, Ceramic

TYPICAL SURFACE MOUNT PC BOARD LAYOUT, ADJUSTABLE OUTPUT (4X SIZE)



 C_{IN} - 15 $\mu\text{F},$ 50V, Solid Tantalum Sprague, "594D series"

 C_{OUT} - 33 $\mu F,$ 25V, Solid Tantalum Sprague, "594D series"

D1 - 1A, 40V Schottky Rectifier, Surface Mount

L1 - 68 μ H, L30, Coilcraft DO3316

C_B - 0.01 μF, 50V, Ceramic

R1 - 1k, 1%

R2 - Use formula in Design Procedure

FIGURE 18. PC Board Layout

Layout is very important in switching regulator designs. Rapidly switching currents associated with wiring inductance can generate voltage transients which can cause problems. For minimal inductance and ground loops, the wires indicated by **heavy lines (in** *Figure 2* and *Figure 3*) should be wide **printed circuit traces and should be kept as short as possible.** For best results, external components should be located as close to the switcher IC as possible using ground plane construction or single point grounding. If **open core inductors are used**, special care must be taken as to the location and positioning of this type of inductor. Allowing the inductor flux to intersect sensitive feedback, IC ground path, and C_{OUT} wiring can cause problems.

When using the adjustable version, special care must be taken as to the location of the feedback resistors and the associated wiring. Physically locate both resistors near the IC, and route the wiring away from the inductor, especially an open core type of inductor.

LLP PACKAGE DEVICES

The LM2672 is offered in the 16 lead LLP surface mount package to allow for increased power dissipation compared to the SO-8 and DIP.

The Die Attach Pad (DAP) can and should be connected to PCB Ground plane/island. For CAD and assembly guidelines refer to Application Note AN-1187 at http:// power.national.com.





Notes

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